

# Quad Picoampere Input Current Bipolar Op Amp

**AD704** 

### FEATURES

High DC Precision
75 μV max Offset Voltage
1 μV/°C max Offset Voltage Drift
150 pA max Input Bias Current
0.2 pA/°C typical I<sub>B</sub> Drift
Low Noise

 $0.5~\mu V$  p-p typical Noise, 0.1 Hz to 10 Hz Low Power

600 μA max Supply Current per Amplifier Chips & MIL-STD-883B Processing Available Available in Tape and Reel in Accordance with EIA-481A Standard

Single Version: AD705, Dual Version: AD706

PRIMARY APPLICATIONS
Industrial/Process Controls
Weigh Scales
ECG/EKG Instrumentation
Low Frequency Active Filters

### PRODUCT DESCRIPTION

The AD704 is a quad, low power bipolar op amp that has the low input bias current of a BiFET amplifier but which offers a significantly lower  $I_B$  drift over temperature. It utilizes Superbeta bipolar input transistors to achieve picoampere input bias current levels (similar to FET input amplifiers at room temperature), while its  $I_B$  typically only increases by 5× at +125°C (unlike a BiFET amp, for which  $I_B$  doubles every 10°C resulting in a 1000× increase at +125°C). Furthermore the AD704 achieves 75  $\mu$ V offset voltage and low noise characteristics of a precision bipolar input op amp.

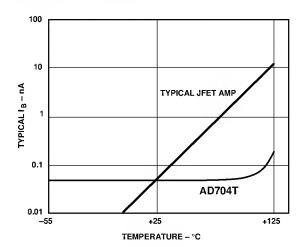


Figure 1. Input Bias Current Over Temperature

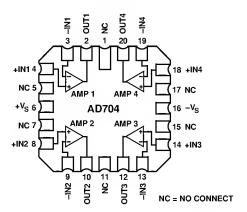
### REV. A

Information furnished by Analog Devices is believed to be accurate and reliable. However, no responsibility is assumed by Analog Devices for its use, nor for any infringements of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of Analog Devices.

#### CONNECTION DIAGRAMS

14-Pin Plastic DIP (N) 16-Pin SOIC 14-Pin Cerdip (Q) Packages (R) Package OUTPUT 16 OUTPUT 13 14 12 **AD704** AD704 13 \_Vs 11 (TOP VIEW) 12 10 +IN 9 10 OUTPUT NC NC = NO CONNECT

### (E) Package 20-Terminal LCC



Since it has only 1/20 the input bias current of an AD OP07, the AD704 does not require the commonly used "balancing" resistor. Furthermore, the current noise is 1/5 that of the AD OP07 which makes the AD704 usable with much higher source impedances. At 1/6 the supply current (per amplifier) of the AD OP07, the AD704 is better suited for today's higher density circuit boards and battery powered applications.

The AD704 is an excellent choice for use in low frequency active filters in 12- and 14-bit data acquisition systems, in precision instrumentation, and as a high quality integrator. The AD704 is internally compensated for unity gain and is available in five performance grades. The AD704J and AD704K are rated over the commercial temperature range of 0°C to +70°C. The AD704A and AD704B are rated over the industrial temperature of -40°C to +85°C. The AD704T is rated over the military temperature range of -55°C to +125°C and is available processed to MIL-STD-883B, Rev. C.

One Technology Way, P.O. Box 9106, Norwood, MA 02062-9106, U.S.A. Tel: 617/329-4700 Fax: 617/326-8703

# AD704—SPECIFICATIONS (@ $T_A = +25^{\circ}C$ , $V_{CM} = 0$ V, and $\pm 15$ V dc, unless otherwise noted)

Model		AD704J/A		AD704K/B			AD704T				
	Conditions	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	Units
INPUT OFFSET VOLTAGE Initial Offset Offset vs. Temp, Average TC vs. Supply (PSRR) $T_{MIN}-T_{MAX}$ Long Term Stability	$T_{MIN}-T_{MAX}$ $V_{S} = \pm 2 \text{ to } \pm 18 \text{ V}$ $V_{S} = \pm 2.5 \text{ to } \pm 18 \text{ V}$	100 100	50 100 0.2 132 126 0.3	150 250 1.5	112 108	30 50 0.2 132 126 0.3	75 150 1.0	112 108	30 80 132 126 0.3	100 150 1.0	μV μV μV/°C dB dB μV/month
INPUT BIAS CURRENT <sup>1</sup> vs. Temp, Average TC $T_{MIN}$ - $T_{MAX}$ $T_{MIN}$ - $T_{MAX}$	$V_{CM} = 0 V$ $V_{CM} = \pm 13.5 V$ $V_{CM} = 0 V$ $V_{CM} = \pm 13.5 V$		100	270 300 300 400		80	150 200 200 300		80	200 250 600 700	pA pA pA/°C pA pA
INPUT OFFSET CURRENT  vs. Temp, Average TC $T_{MIN}$ $T_{MAX}$ $T_{MIN}$ $T_{MAX}$	$V_{CM} = 0 V$ $V_{CM} = \pm 13.5 V$ $V_{CM} = 0 V$ $V_{CM} = \pm 13.5 V$		80 0.6 100 100	250 300 300 400		30 0.4 80 80	100 150 200 300		50 0.4 80 100	150 200 400 500	pA pA pA/°C pA pA
MATCHING CHARACTERISTICS Offset Voltage  Input Bias Current <sup>2</sup> Common-Mode Rejection <sup>3</sup> Power Supply Rejection <sup>4</sup> Crosstalk <sup>5</sup>	$T_{MIN}-T_{MAX}$ $T_{MIN}-T_{MAX}$ $T_{MIN}-T_{MAX}$ $T_{MIN}-T_{MAX}$ $f = 10 \text{ Hz}$	94 94 94 94		250 400 500 600	110 104 110 106		130 200 300 400	104 104 110 106		150 250 400 600	μV μV pA pA dB dB dB dB
FREQUENCY RESPONSE UNITY GAIN Crossover Frequency Slew Rate, Unity Gain Slew Rate	$R_{LOAD} = 2 \text{ k}\Omega$ $G = -1$ $T_{MIN} - T_{MAX}$		0.8 0.15 0.1			0.8 0.15 0.1			0.8 0.15 0.1		MHz V/µs V/µs
INPUT IMPEDANCE Differential Common-Mode			40  2 300  2			40  2 300  2			40  2 300  2		$M\Omega \  pF$ $G\Omega \  pF$
INPUT VOLTAGE RANGE Common-Mode Voltage Common-Mode Rejection Ratio	$V_{CM} = \pm 13.5 \text{ V}$ $T_{MIN} - T_{MAX}$	±13.5 100 98	±14 132 128		±13.5 114 108	±14 132 128		±13.5 110 108	±14 132 128		V dB dB
INPUT CURRENT NOISE	0.1 to 10 Hz f = 10 Hz		3 50			3 50			3 50		pA p-p fA/√Hz
INPUT VOLTAGE NOISE	0.1 to 10 Hz f = 10 Hz f = 1 kHz		0.5 17 15	22		0.5 17 15	2.0 22		0.5 17 15	2.0 22	$\begin{array}{c} \mu V \ p-p \\ nV/\sqrt{Hz} \\ nV/\sqrt{Hz} \end{array}$
OPEN-LOOP GAIN	$\begin{aligned} V_O &= \pm 12 \text{ V} \\ R_{LOAD} &= 10 \text{ k}\Omega \\ T_{MIN} - T_{MAX} \\ V_O &= \pm 10 \text{ V} \\ R_{LOAD} &= 2 \text{ k}\Omega \\ T_{MIN} - T_{MAX} \end{aligned}$	200 150 200 150	2000 1500 1000 1000		400 300 300 200	2000 1500 1000 1000		400 300 200 100	2000 1500 1000 1000		V/mV V/mV V/mV V/mV

-2- REV. A

Model		AD704J/A			AD704K/B			AD704T			
	Conditions	Min	Typ	Max	Min	Typ	Max	Min	Typ	Max	Units
OUTPUT CHARACTERISTICS											
Voltage Swing	$R_{LOAD} = 10 \text{ k}\Omega$										
	$T_{MIN}-T_{MAX}$	±13	$\pm 14$		± 13	$\pm 14$		±13	$\pm 14$		V
Current	Short Circuit		±15			±15			±15		mA
CAPACITIVE LOAD											
Drive Capability	Gain = + 1		10,00	0		10,00	00		10,00	0	pF
POWER SUPPLY											
Rated Performance			±15			±15			±15		V
Operating Range		±2.0		$\pm 18$	±2.0		$\pm 18$	±2.0		$\pm 18$	V
Quiescent Current			1.5	2.4		1.5	2.4		1.5	2.4	mA
	$T_{MIN}-T_{MAX}$		1.6	2.6		1.6	2.6		1.6	2.6	mA
TRANSISTOR COUNT	# of Transistors		180			180			180		

#### NOTES

### ABSOLUTE MAXIMUM RATINGS<sup>1</sup>

Supply Voltage
Internal Power Dissipation (+25°C) See Note 2
Input Voltage
Differential Input Voltage <sup>3</sup>
Output Short Circuit Duration (Single Input) Indefinite
Storage Temperature Range
(Q)65°C to +150°C

(Q)	-05 C to +150 C
(N, R)	-65°C to +125°C
Operating Temperature Range	

Lead Temperature Range (Soldering 10 seconds) .... +300°C

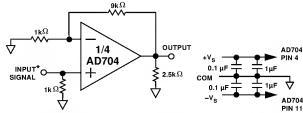
### NOTES

<sup>1</sup>Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

<sup>2</sup>Specification is for device in free air:

14-Pin Plastic Package:  $\theta_{JA} = 150^{\circ}\text{C/Watt}$ 14-Pin Cerdip Package:  $\theta_{JA} = 110^{\circ}\text{C/Watt}$ 16-Pin SOIC Package:  $\theta_{JA} = 100^{\circ}\text{C/Watt}$ 20-Terminal LCC Package:  $\theta_{JA} = 150^{\circ}\text{C/Watt}$ 

<sup>3</sup>The input pins of this amplifier are protected by back-to-back diodes. If the differential voltage exceeds ±0.7 volts, external series protection resistors should be added to limit the input current to less than 25 mA.

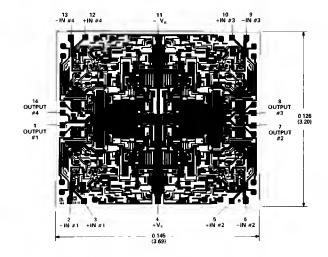


ALL 4 AMPLIFIERS ARE CONNECTED AS SHOWN

### Figure 2a. Crosstalk Test Circuit

### METALIZATION PHOTOGRAPH

Dimensions shown in inches and (mm). Contact factory for latest dimensions.



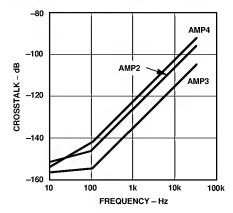


Figure 2b. Crosstalk vs. Frequency

<sup>&</sup>lt;sup>1</sup>Bias current specifications are guaranteed maximum at either input.

<sup>&</sup>lt;sup>2</sup>Input bias current match is the maximum difference between corresponding inputs of all four amplifiers.

 $<sup>^3</sup>$ CMRR match is the difference of  $\Delta V_{OS}/\Delta V_{CM}$  between any two amplifiers, expressed in dB.

<sup>&</sup>lt;sup>4</sup>PSRR match is the difference between  $\Delta V_{OS}/\Delta V_{SUPPLY}$  for any two amplifiers, expressed in dB.

<sup>&</sup>lt;sup>5</sup>See Figure 2a for test circuit.

All min and max specifications are guaranteed.

Specifications subject to change without notice.

<sup>\*</sup>THE SIGNAL INPUT (SUCH THAT THE AMPLIFIER'S OUTPUT IS AT MAX AMPLITUDE WITHOUT CLIPPING OR SLEW LIMITING) IS APPLIED TO ONE AMPLIFIER AT A TIME. THE OUTPUTS OF THE OTHER THREE AMPLIFIERS ARE THEN MEASURED FOR CROSSTALK.

# AD704—Typical Characteristics (@ $\pm 25^{\circ}$ C, $V_s = \pm 15$ V, unless otherwise noted)

### **ORDERING GUIDE**

Model	Temperature Range	Package Option*
AD704JN	0°C to +70°C	N-14
AD704JR	0°C to +70°C	R-16
AD704JR-/REEL	0°C to +70°C	Tape and Reel
AD704KN	0°C to +70°C	N-14
AD704AN	−40°C to +85°C	N-14
AD704AQ	−40°C to +85°C	Q-14
AD704AR	−40°C to +85°C	R-16
AD704AR-REEL	−40°C to +85°C	Tape and Reel
AD704BQ	−40°C to +85°C	Q-14
AD704SE/883B	−55°C to +125°C	E-20A
AD704TQ	−55°C to +125°C	Q-14
AD704TQ/883B	-55°C to +125°C	Q-14

Chips are also available.

<sup>\*</sup>E = Leadless Ceramic Chip Carrier; N = Plastic DIP; Q = Cerdip; R = Small Outline (SOIC).

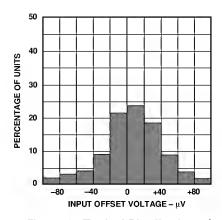


Figure 3. Typical Distribution of Input Offset Voltage

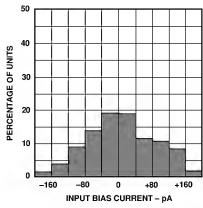


Figure 4. Typical Distribution of Input Bias Current

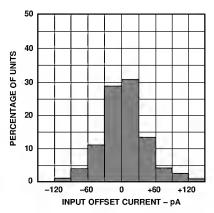


Figure 5. Typical Distribution of Input Offset Current

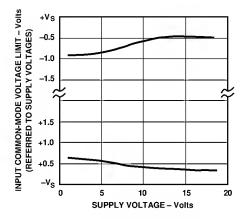


Figure 6. Input Common-Mode Voltage Range vs. Supply Voltage

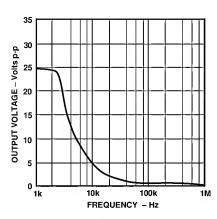


Figure 7. Large Signal Frequency Response

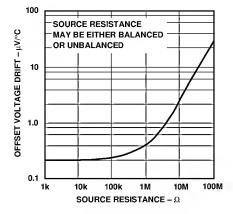


Figure 8. Offset Voltage Drift vs. Source Resistance

4– REV. A

# AD704

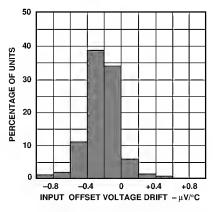


Figure 9. Typical Distribution of Offset Voltage Drift

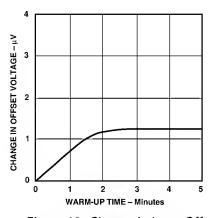


Figure 10. Change in Input Offset Voltage vs. Warm-Up Time

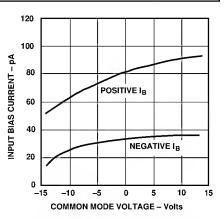


Figure 11. Input Bias Current vs. Common-Mode Voltage

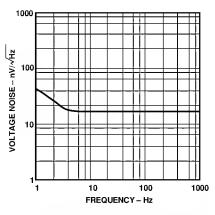


Figure 12. Input Noise Voltage Spectral Density

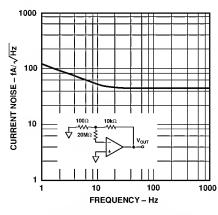


Figure 13. Input Noise Current Spectral Density

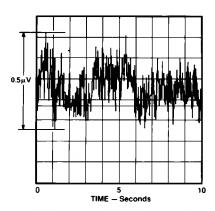


Figure 14. 0.1 Hz to 10 Hz Noise Voltage

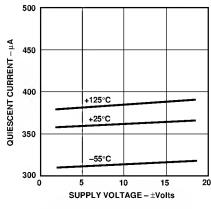


Figure 15. Quiescent Supply Current vs. Supply Voltage (per Amplifier)

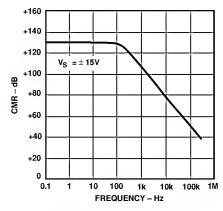


Figure 16. Common-Mode Rejection vs. Frequency

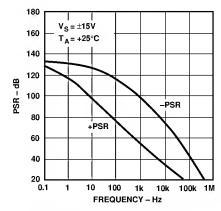


Figure 17. Power Supply Rejection vs. Frequency

REV. A -5-

## AD704

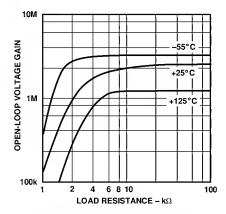


Figure 18. Open-Loop Gain vs. Load Resistance Over Temperature

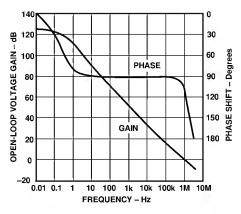


Figure 19. Open-Loop Gain and Phase vs. Frequency

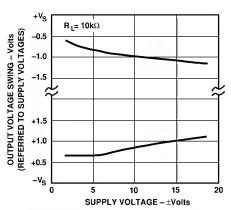


Figure 20. Output Voltage Swing vs. Supply Voltage

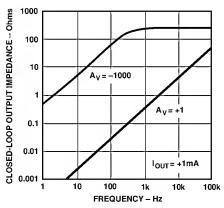


Figure 21. Closed-Loop Output Impedance vs. Frequency

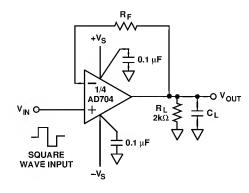


Figure 22a. Unity Gain Follower (For Large Signal Applications, Resistor  $R_F$  Limits the Current Through the Input Protection Diodes)

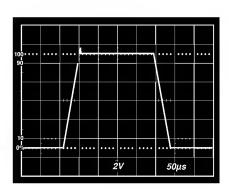


Figure 22b. Unity Gain Follower Large Signal Pulse Response  $R_F = 10 \text{ k}\Omega$ ,  $C_L = 1,000 \text{ pF}$ 

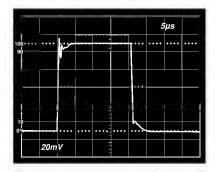


Figure 22c. Unity Gain Follower Small Signal Pulse Response  $R_F = 0 \Omega$ ,  $C_L = 100 pF$ 

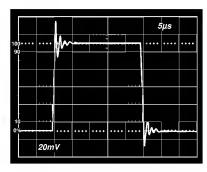


Figure 22d. Unity Gain Follower Small Signal Pulse Response  $R_F = 0 \Omega$ ,  $C_L = 1,000 pF$ 

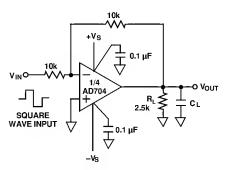


Figure 23a. Unity Gain Inverter Connection

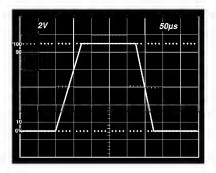


Figure 23b. Unity Gain Inverter Large Signal Pulse Response, C<sub>I</sub> = 1,000 pF

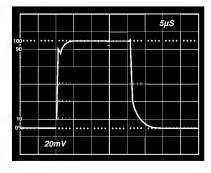


Figure 23c. Unity Gain Inverter Small Signal Pulse Response,  $C_L = 100 \text{ pF}$ 

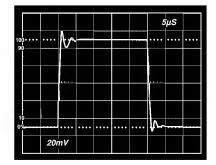


Figure 23d. Unity Gain Inverter Small Signal Pulse Response,  $C_L = 1,000 \text{ pF}$ 

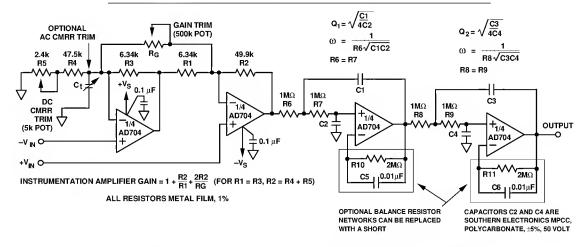


Figure 24. Gain of 10 Instrumentation Amplifier with Post Filtering

The instrumentation amplifier with post filtering (Figure 24) combines two applications which benefit greatly from the AD704. This circuit achieves low power and dc precision over temperature with a minimum of components.

The instrumentation amplifier circuit offers many performance benefits including BiFET level input bias currents, low input offset voltage drift and only 1.2 mA quiescent current. It will operate for gains  $G \ge 2$ , and at lower gains it will benefit from the fact that there is no output amplifier offset and noise contribution as encountered in a 3 op amp design. Good low frequency CMRR is achieved even without the optional AC CMRR trim (Figure 25). Table I provides resistance values for 3 common circuit gains. For other gains, use the following equations:

$$R2 = R4 + R5 = 49.9 \ k\Omega$$

$$R1 = R3 = \frac{49.9 \ k\Omega}{0.9 \ G - 1}$$

Max Value of 
$$R_G = \frac{99.8 k}{0.06 G}$$

$$C_t \approx \frac{1}{2 \pi (R3) 5 \times 10^5}$$

Table I. Resistance Values for Various Gains

Circuit Gain (G)	R1 & R3	R <sub>G</sub> (Max Value of Trim Potentiometer)	Bandwidth (-3 dB), Hz
10	6.34 kΩ	166 kΩ	50k
100	$526~\Omega$	16.6 kΩ	5k
1,000	$56.2~\Omega$	1.66 kΩ	0.5k

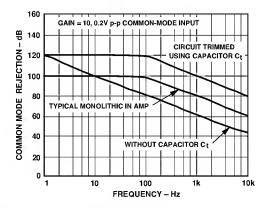


Figure 25. Common-Mode Rejection vs. Frequency with and without Capacitor  $C_{\rm t}$ 

### AD704

The 1 Hz, 4-pole active filter offers dc precision with a minimum of components and cost. The low current noise,  $I_{OS}$ , and  $I_B$  allow the use of 1  $M\Omega$  resistors without sacrificing the 1  $\mu V/^{\circ} C$  drift of the AD704. This means lower capacitor values may be used, reducing cost and space. Furthermore, since the AD704's  $I_B$  is as low as its  $I_{OS}$ , over most of the MIL temperature range, most applications do not require the use of the normal balancing resistor (with its stability capacitor). Adding the optional balancing resistor enhances performance at high temperatures, as shown in Figure 26. Table II gives capacitor values for several common low pass responses.

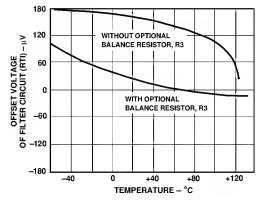


Figure 26.  $V_{OS}$  vs. Temperature Performance of the 1 Hz Filter Circuit

Table II. 1 Hz, 4-Pole Low-Pass Filter Recommended Component Values

Desired Low Pass Response	Section 1 Frequency (Hz)	Q	Section 2 Frequency (Hz)	Q	C1 (µF)	C2 (µF)	C3 (μF)	C4 (μF)
Bessel	1.43	0.522	1.60	0.806	0.116	0.107	0.160	0.0616
Butterworth	1.00	0.541	1.00	1.31	0.172	0.147	0.416	0.0609
0.1 dB Chebychev	0.648	0.619	0.948	2.18	0.304	0.198	0.733	0.0385
0.2 dB Chebychev	0.603	0.646	0.941	2.44	0.341	0.204	0.823	0.0347
0.5 dB Chebychev	0.540	0.705	0.932	2.94	0.416	0.209	1.00	0.0290
1.0 dB Chebychev	0.492	0.785	0.925	3.56	0.508	0.206	1.23	0.0242

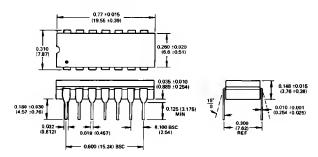
Specified Values are for a -3 dB point of 1.0 Hz. For other frequencies simply scale capacitors C1 through C4 directly; i.e., for 3 Hz Bessel response, C1 = 0.0387  $\mu$ F, C2 = 0.0357  $\mu$ F, C3 = 0.0533  $\mu$ F, C4 = 0.0205  $\mu$ F.

### **OUTLINE DIMENSIONS**

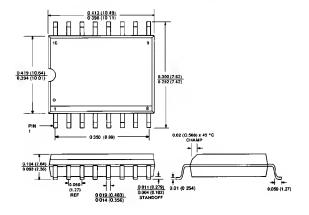
Dimensions shown in inches and (mm).

-8-

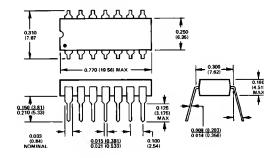
### 14-Pin Cerdip (Q) Package



### 16-Pin Plastic SO (R) Package



### 14-Pin Plastic DIP (N) Package



### 20-Terminal LCCC (E) Package

